

Expanding the Field Range of PHE Sensors for Increased Industrial Applicability

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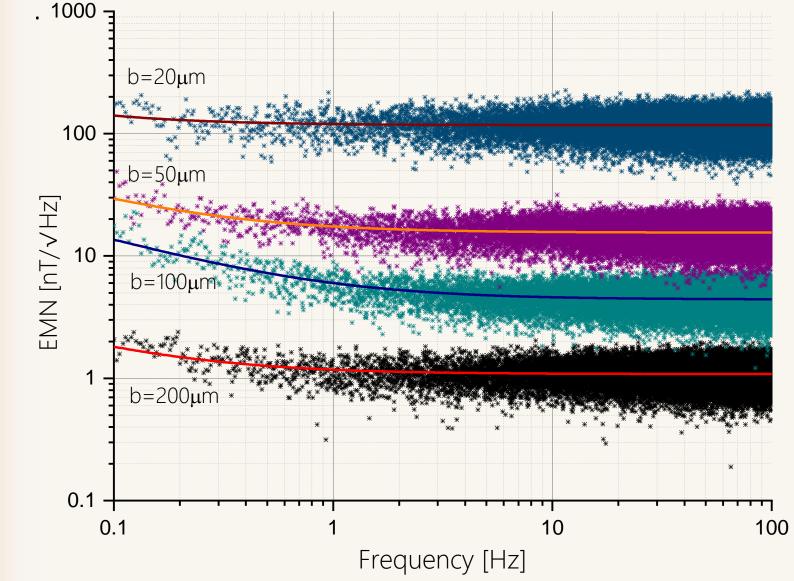
Motivation - The Need for Range

- Magnetic field sensing plays a pivotal role in numerous technological applications, including automotive, biomedical diagnostics and more.
- Planar Hall effect (PHE) sensors provide superior performance, simpler fabrication, high resolution, and low temperature dependence compared to other sensors.
- The field range of high-resolution PHE sensors is typically limited to hundreds of micro-Tesla, restricting their applicability in certain critical areas.
- Extending the field range of PHE sensors is essential to meet the demand for measuring a wide range of field strengths in industrial, scientific, and consumer applications. This extension **enhances applicability** while requiring careful evaluation of its impact on performance metrics, such as equivalent magnetic noise (EMN), to optimize sensor

performance.

Results

Table 1 demonstrates that reducing the hard axis dimension (b) while maintaining the a/b ratio extends the field range of the EPHE sensor. At a constant current density, the 20-micron sensor's sensitivity is approximately 100 times lower than the 200-micron sensor, in close agreement with theoretical predictions.



b (μm)	I (mA)	H_{eff} (Oe)	$\frac{S_y}{(mV/T)}$	$\Delta R_{\rm PHE}$ (m Ω)	R_x (Ω)	R_y (Ω)
20	3	124.8	7.2	29	8.9	35.6
50	7.5	43.5	46	26	7	20.7
100	15	23.8	151.5	25	5.6	14.4
200	30	13.5	642.1	26	5.1	9.8

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Table 1. Typical values of $H_{\rm eff}$, S_y , $\Delta R_{\rm PHE}$, R_x , and R_y for EPHE sensors with varying hard axis lengths, while maintaining uniform current density.

Figure 1 shows the EMN from 0.1 to 100 Hz for sensors with varying hard axis values. The EMN is fitted using





Fabrication Process

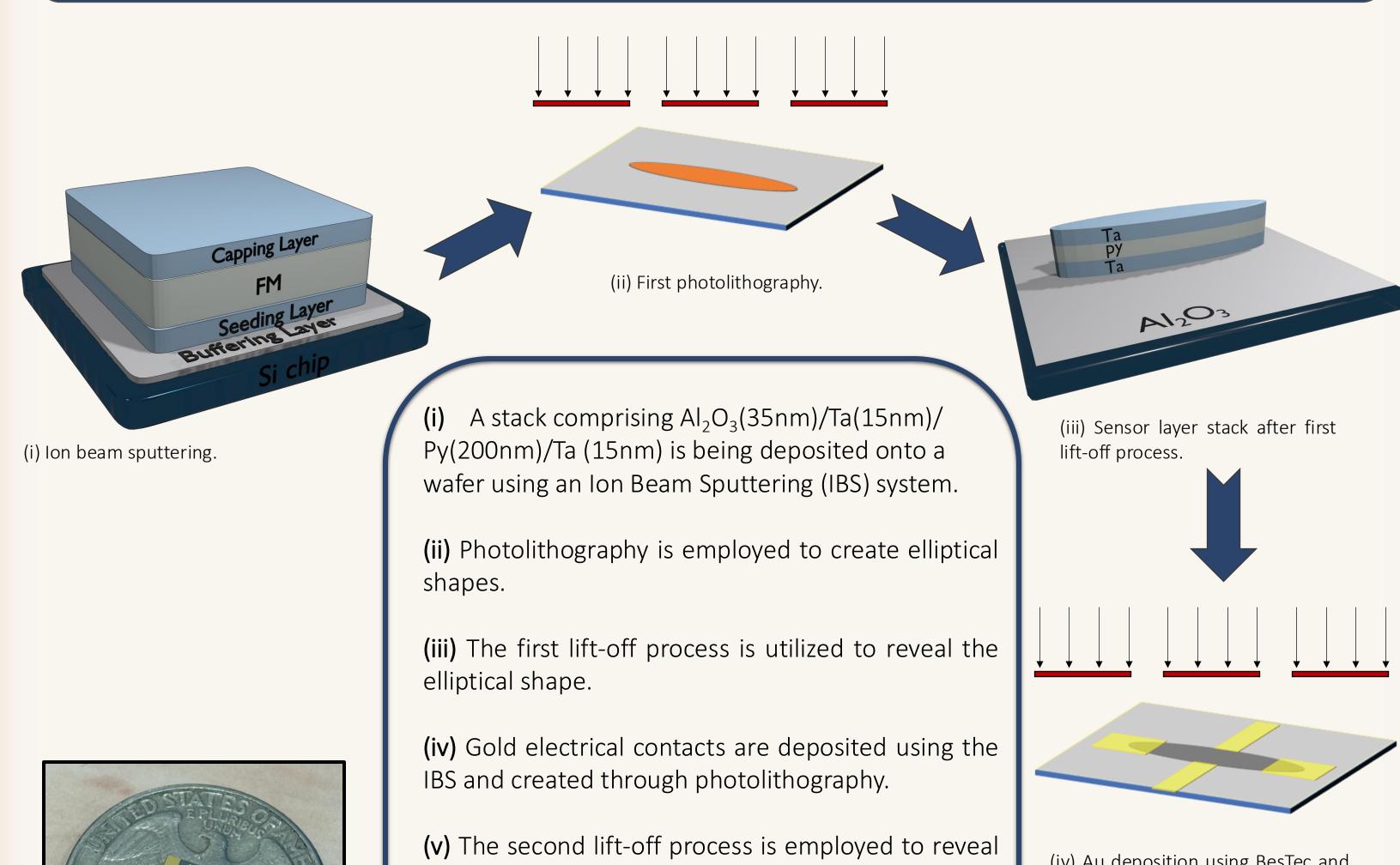


Figure 1. Equivalent magnetic noise (EMN) of the PHE sensors versus frequency for different hard axis (b) values, with the same current density.

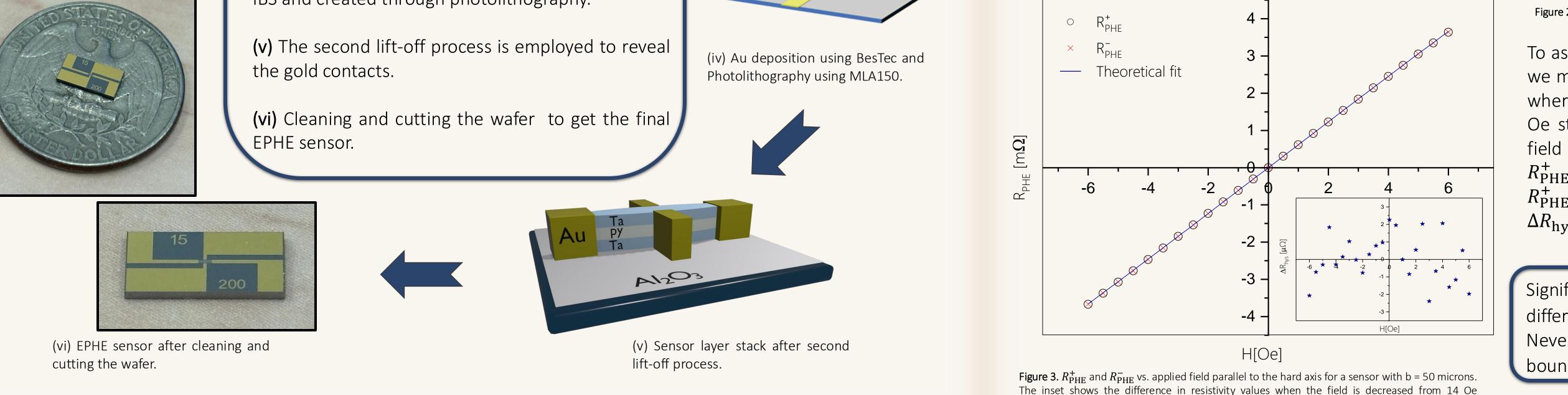
Figure 2 shows the extended field range of the sensors in response to a magnetic field along the hard axis. The resistivity was measured during both ascending and descending fields and compared to the theoretical fit from the Stoner–Wohlfarth (SW) model:

$$R_{\rm PHE} = \frac{\Delta R_{\rm PHE}}{H_{\rm eff}} \cdot H_{\sqrt{1 - \left(\frac{H}{H_{\rm eff}}\right)^2}}$$

The inset presents the scaled data with normalized values $R_{\text{PHE}}^* = \frac{R_{\text{PHE}}}{\Delta R_{\text{PHE}}}$ and $H^* = \frac{H}{H_{\text{eff}}}$.

> It is evident that although all sensors exhibit non-linear behavior, this behavior can be precisely modeled, which means that they effectively behave as single magnetic domains.

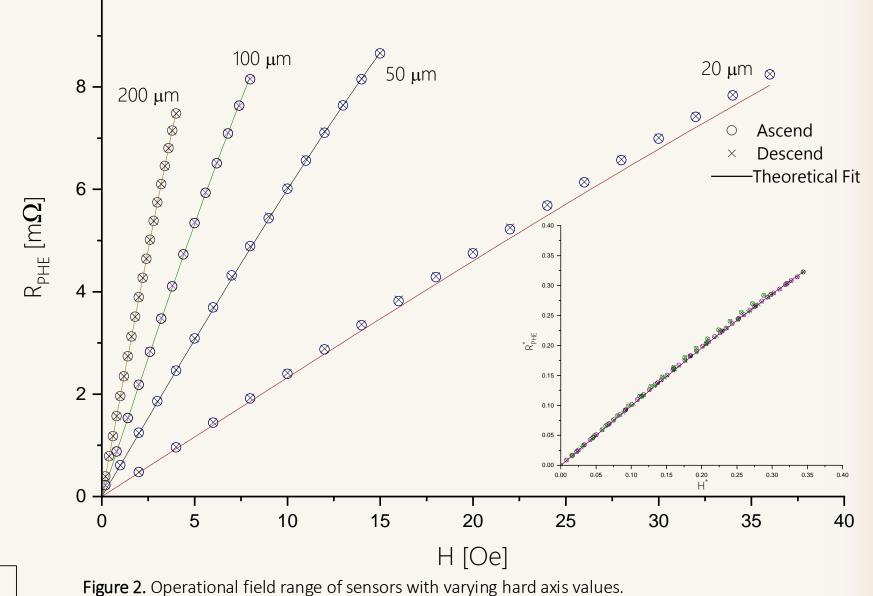
compared to when it is increased from -14 Oe for the same sensor.



 $EMN(f) = \sqrt{B^2 + \frac{A^2}{f}}$, where A and B are fitting

parameters. At 10 Hz, the EMNs are approximately 1.1, 4.4, 16, and 117 nT/\sqrt{Hz} for hard axis values of 200, 100, 50, and 20 microns, respectively. The results demonstrate that the EMN scales roughly as n^2 , with the 20-micron sensor having an EMN about 100 times larger than the 200-micron sensor, as predicted.

Notably, doubling the current density results in EMNs of 56 $nT/_{\sqrt{Hz}}$ and 7.7 $nT/_{\sqrt{Hz}}$ at 10 Hz for the 20-micron and 50-micron sensors, respectively.



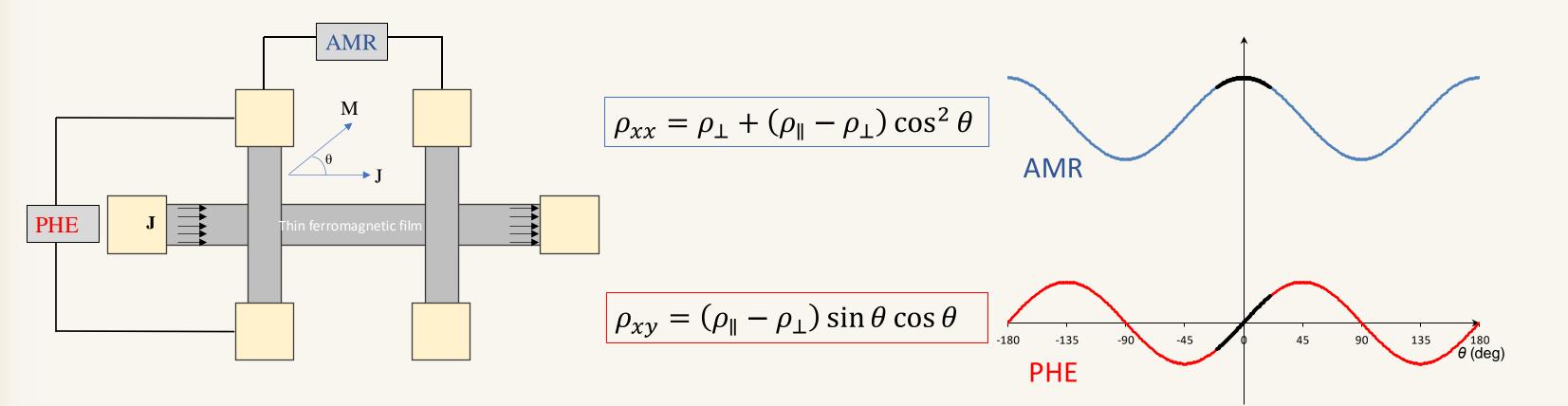
To assess the hysteresis of a sensor with b = 50 microns, we measured $R_{\rm PHE}$ in a field range from -6 Oe to +6 Oe, where deviation from linearity is less than 1%. Using 0.5 Oe steps, $R_{\rm PHE}$ was measured twice: first reducing the field from +14 Oe and then increasing it from -14 Oe. $R_{\rm PHE}^+$ and $R_{\rm PHE}^-$ represent the averages. Figure 3 shows $R_{\rm PHE}^+$ and $R_{\rm PHE}^-$, while the inset presents the difference, $\Delta R_{\rm hys} = R_{\rm PHE}^+ - R_{\rm PHE}^-.$

Significantly, $\Delta R_{
m hys}$ changes sign, indicating that the differences are largely unrelated to magnetic hysteresis. Nevertheless, ΔR_{hys} can still be considered an upper bound for any potential magnetic hysteresis.

Principle and Design

Planar Hall Effect (PHE)

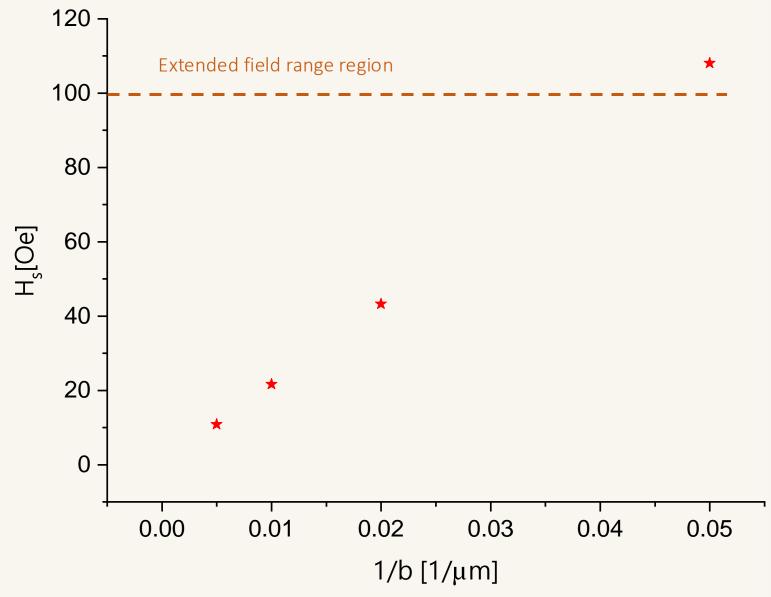
Ferromagnetic materials exhibit a phenomenon called anisotropic magnetoresistance (AMR), in which their resistance changes depending on the orientation of their magnetization (\mathbf{M}) relative to the direction of an applied electric current density (\mathbf{J}) . This phenomenon gives rise to a transverse voltage as a function of the angle θ between **M** and **J**. This effect is called PHE as the magnetization, the electric current, and the transverse electric field are in the same plane.



A Novel EPHE Sensor with Extended Field Range

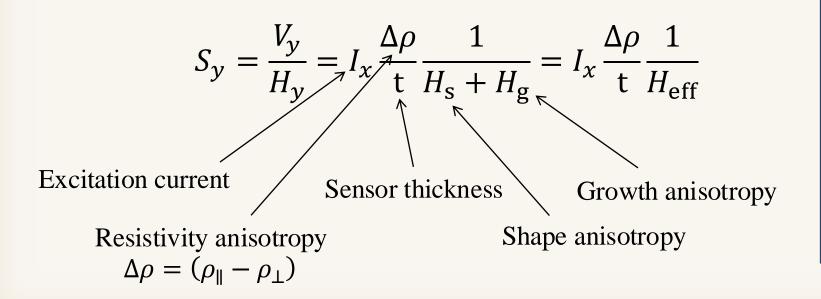
The shape of the elliptical PHE (EPHE) sensors induces a uniaxial magnetic anisotropy parallel to the long axis (the easy axis), known as "shape anisotropy". For an elongated and relatively flat ellipsoid (thickness t) with principal axes a and b $(a \gg b \gg t)$ will be:

$$H_{\rm s} \sim 4\pi M_{\rm s} \frac{t}{t} \sim 10807 \frac{t}{t}$$



Sensitivity

The sensitivity (S_{γ}) of an EPHE sensor is the ratio of its PHE voltage (V_y) to the applied magnetic field in the y direction (H_{γ}) , for a given current I_{χ} along the easy axis. When H_{γ} is small relative to the effective anisotropy field (H_{eff}) , the sensitivity is given by

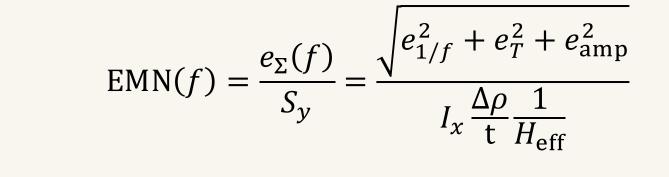


Equivalent Magnetic Noise (EMN)

The total noise, e_{Σ} , has three main components: 1/f noise, thermal noise, and preamplifier noise.

$$e_{\Sigma}(f) = \sqrt{e_{1/f}^2 + e_T^2 + e_{amp}^2}$$

The sensor EMN is defined as



Utilizing this relation, we computed theoretical values of the shape anisotropy field for a sensor featuring a thickness of 200 nm across various hard axis lengths.

> A sensor with a hard axis length of 20 microns surpasses 100 Oe, qualifying it as an extended field range sensor.

Design

We fabricate EPHE sensors with a thickness of 200 nm and hard axis lengths of 200, 100, 50, and 20 microns.

A schematic diagram of the sensor with the ellipse's major and minor axes (denoted as a and b, respectively), with aspect ratio 1:8, and the placement of the gold electrical contact pads (highlighted in yellow).

The sensor is excited by applying an ac current between V_{x1} and V_{x2} . The signal is measured between V_{y1} and V_{y2} .